

CHAPTER 30: INDUCTANCE

Inductance = coupling of conducting coils through
changing magnetic field or flux



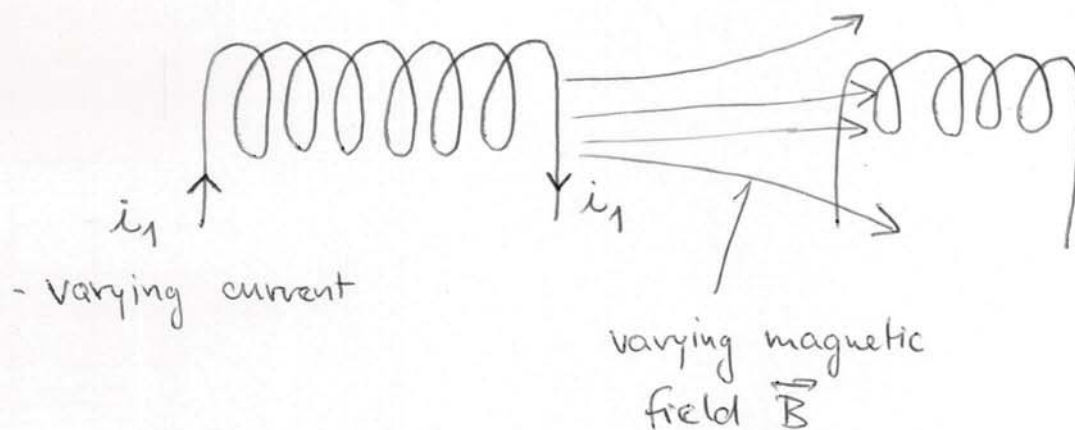
Coils are inductors

Section 30.1: Mutual Inductance

- Consider two neighboring conducting coils

coil 1 - N_1 turns

coil 2 - N_2 turns



- varying current

varying magnetic
field $\vec{B}$

- current i_1 flowing through coil 1 produces magnetic field $\vec{B}$ and magnetic flux Φ_{B2} through coil 2



if the current i_1 changes, the flux Φ_{B2} also changes



electromotive force (emf) is induced in coil 2,

Following Faraday's law

- If Φ_{B2} denotes average magnetic flux through coil 2,

Faraday's law gives the induced emf as:

$$\boxed{\mathcal{E}_2 = - N_2 \frac{d\Phi_{B2}}{dt}}$$

- Let us now define the mutual inductance of coils 1 and 2

as:

$$\boxed{M_{21} = \frac{N_2 \Phi_{B2}}{i_1} \Rightarrow M_{21} i_1 = N_2 \Phi_{B2}}$$

M_{21} - constant of proportionality

↓

With this definition of M_{21} , we can rewrite Faraday's

law as:

$$\boxed{\mathcal{E}_2 = - M_{21} \frac{di_1}{dt}}$$

→ induced emf in coil 2 is directly linked to the change of current in coil 1

M_{21} - a constant depending on the geometry of both coils (size, shape, number of turns), orientation and separation of coils, and magnetic properties of material between the coils

- for linear magnetic response of the material between the coils, M_{21} is independent of i_1

- If we reverse the experiment $\rightarrow$
 changing current in coil 2 induces emf in coil 1
 $\rightarrow$ experiments show that $M_{12} = M_{21}$ for
 all configurations of coils 1 and 2
 $\rightarrow$ common mutual inductance M characterizing
 induced-emf interactions of two coils

$$\boxed{\mathcal{E}_2 = -M \frac{di_1}{dt} \quad \text{and} \quad \mathcal{E}_1 = -M \frac{di_2}{dt}}$$

where

$$\boxed{M = \frac{N_2 \Phi_{B2}}{i_1} = \frac{N_1 \Phi_{B1}}{i_2}}$$

- Negative sign in induced emf equation reflects
 Lenz's law

- Only time-varying current induces an emf !!!

Units of M : $\text{Henry} \equiv 1 \text{ H} = 1 \frac{\text{Wb}}{\text{A}} = 1 \frac{\text{J}}{\text{A}^2}$

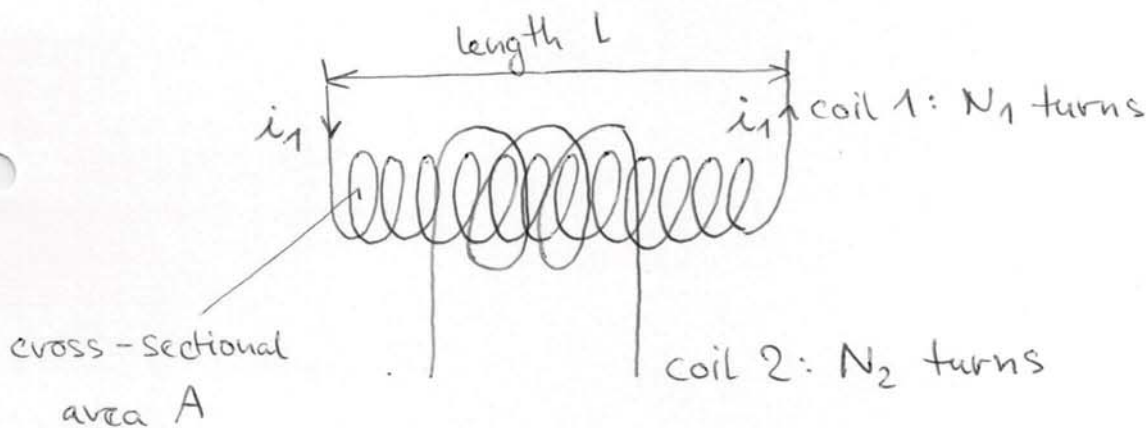
- Mutual inductance can lead to unwanted coupling
 in electric circuits

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mutual inductance is useful in transformers for
 stepping up/down the voltage

Ex 30.1: Calculating mutual inductance

A long solenoid with length L and cross-sectional area A is closely wound with N_1 turns of wire. A coil with N_2 turns surrounds it at its center. Find the mutual inductance



- Magnetic field $\vec{B}_1$ of a solenoid close to its center is uniform and points along the axis of the solenoid:

$$B_1 = \mu_0 n_1 i_1 = \mu_0 \frac{N_1}{L} i_1$$

$i_1 \dots$ current through the solenoid

$n_1 \dots$ number of solenoid turns per unit length

- Flux through the cross-section of the solenoid: $B_1 A$



Since there is no field outside solenoid 1, we can write

for the flux through solenoid 2 $\Phi_{B2} = B_1 \cdot A$

• From definition

$$\begin{aligned}
 M &= \frac{N_2 \Phi_{B2}}{i_1} = \\
 &= \frac{N_2 B_1 A}{i_1} = \frac{N_2}{i_1} \frac{\mu_0 N_1 i_1}{L} \cdot A = \\
 &= \frac{\mu_0 A N_1 N_2}{L}
 \end{aligned}$$

→ The mutual inductance of two coils is proportional to the product $N_1 N_2$ of their turns.

It is independent of the current.

Section 30.2: Self-Inductance and Inductors

- ② What happens in a single isolated circuit with changing current?



The current sets up a magnetic field that causes a magnetic flux through the same circuit.

This flux changes when the current changes



Emf is induced in the circuit by the variation of its own magnetic field → self-induced emf

- From Lenz's law, self-induced emf always opposes the change in the current that caused the emf
→ current variations are more difficult!
- Self-induced emfs can occur in any current-carrying circuit
→ it is especially pronounced in coils with N turns of wire
- We can define self-inductance L as:

$$L = \frac{N \Phi_B}{i}$$

Φ_B - average magnetic flux through each turn of the coil

i - current through the coil

- Units of self-inductance are identical to the units of mutual inductance $\rightarrow 1 \text{ henry} \equiv 1 \text{ H}$
- Using definition of self-inductance, we can rearrange Faraday's law as:

$$\boxed{\mathcal{E} = -N \frac{d\Phi_B}{dt} = -L \frac{di}{dt}}$$

$\rightarrow$ negative sign reflects Lenz's law
 $\Downarrow$
 induced emf opposes any current changes in the circuit

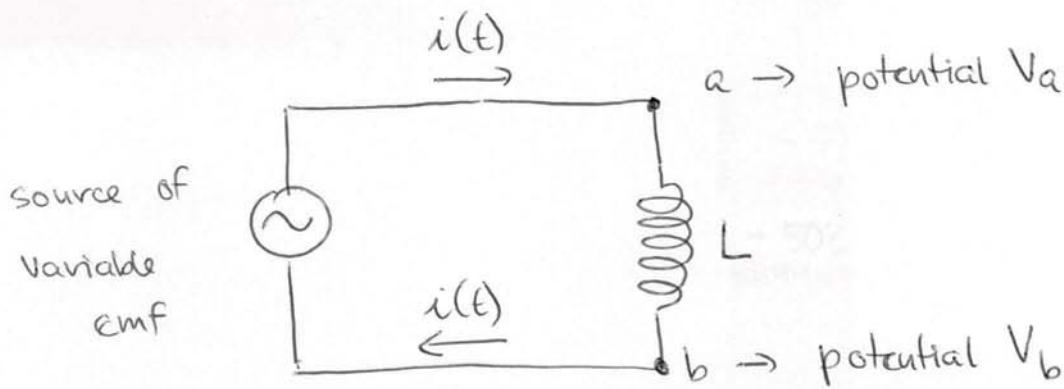
- Faraday's law can be used to measure unknown self-inductance L from a known rate di/dt and measured induced emf $\mathcal{E}$

Inductors as circuit elements

Inductor ... device with a particular (self-) inductance

symbol 

- Inductors are used to oppose any variations in the current through the circuit
 $\rightarrow$ current stabilization



- Due to induced emf, charge accumulates on the terminals of the inductor

→ this generates potential difference $V_{ab} = V_a - V_b$ across the inductor

$$\boxed{V_{ab} = L \frac{di}{dt}} \rightarrow \text{non-conservative induced emf produces conservative electric field with potential}$$

→ We can apply Kirchhoff's loop rule to analyze currents and voltage drops in circuits containing inductors

Potential difference across resistor vs. inductor

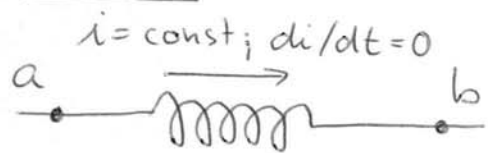
Resistor



$$V_{ab} = iR > 0$$

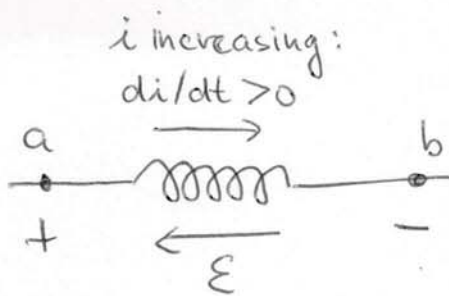
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Inductor

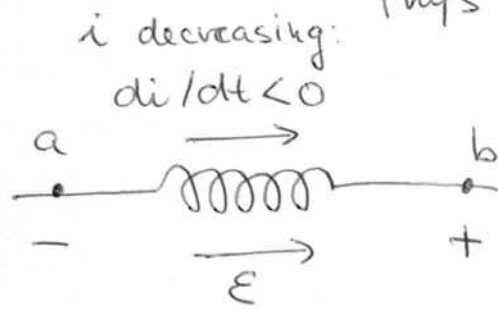


$$\mathcal{E} = 0$$

$$V_{ab} = L \frac{di}{dt} = 0$$



$$V_{ab} = L \frac{di}{dt} > 0$$



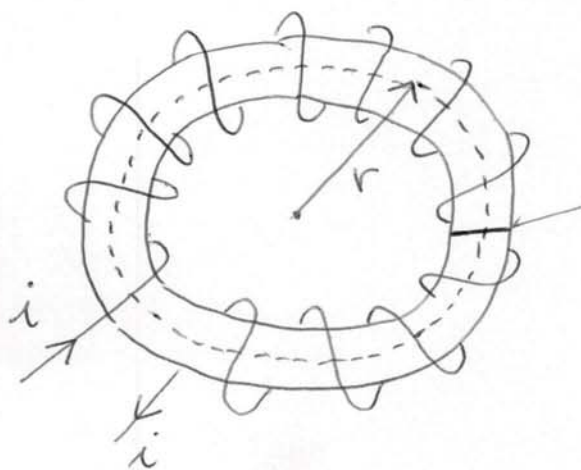
$$V_{ab} = L \frac{di}{dt} < 0$$

Ex 30.3 Calculating self-inductance

A toroidal solenoid with cross-sectional area A and mean radius r is closely wound with N turns of wire. The toroid is wound on a nonmagnetic core.

Determine its self-inductance L , assuming that magnetic field B is uniform across a cross-section.

number of turns = N



cross-sectional area A

assumption of uniform field

• From definition, self-inductance $L = \frac{N\Phi_B}{i} \approx \frac{N \cdot A \cdot B}{i}$

• Magnetic field at a distance r from the toroid axis is calculated from Ampere's law

$$\oint \vec{B} \cdot d\vec{l} = \mu_0 I_{\text{encl}}$$

$$I_{\text{encl}} = N \cdot i$$

$\vec{B}$ is tangential to the toroid $\rightarrow \oint \vec{B} d\vec{l} = \oint B dl =$
 $= B \oint dl = 2\pi r B$

$\Downarrow$

$$2\pi r B = \mu_0 N i \rightarrow \underline{\underline{B = \frac{\mu_0 N i}{2\pi r}}}$$

• Inserting B into the definition of self-inductance

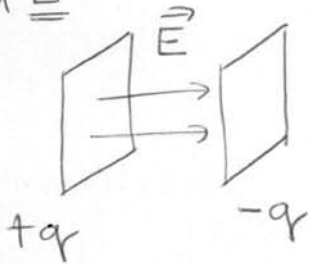
gives
$$L = \frac{N \cdot A \cdot B}{i} = \frac{NA}{i} \frac{\mu_0 N i}{2\pi r} = \underline{\underline{\frac{\mu_0 A N^2}{2\pi r}}}$$

$\rightarrow$ Self-inductance is proportional to the square
 of the number of turns in the coil
 (for closely wound coils)

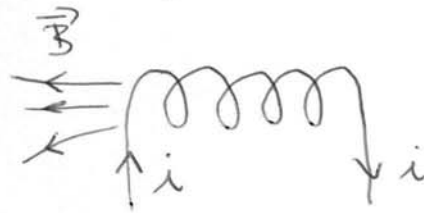
Section 30.3: Magnetic-Field Energy

- An inductor carrying a current has energy stored in it

In a capacitor, stored energy is used to build up electric field $\vec{E}$



In an inductor, stored energy is used to build up magnetic field $\vec{B}$

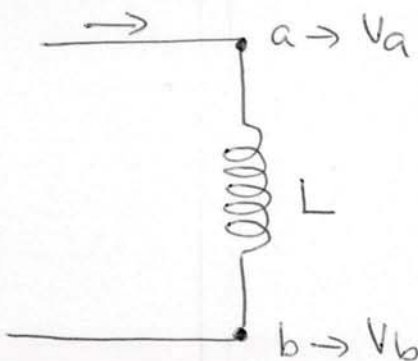


Energy stored in an inductor

- ② How much energy is needed to establish a final current $\underline{I}$ in an inductor with inductance $\underline{L}$ if the initial current is zero?

- Let us assume an inductor with zero resistance
 $\rightarrow$ Zero energy dissipation

$i(t) \equiv i \rightarrow$ time-varying current with rate $\frac{di}{dt} > 0$



$\downarrow$
Voltage drop across the inductor

$$V_{ab} = L \frac{di}{dt}$$

$\downarrow$
Power is delivered to the inductor at the rate $P = V_{ab} \cdot i = \underline{\underline{L i \frac{di}{dt}}}$

- Energy supplied to the inductor in time dt is $dU = P \cdot dt = \underline{\underline{L i di}}$



The total energy U supplied while the current increases from zero to the final value I is

$$U = \int_0^I dU = L \int_0^I i di = \frac{LI^2}{2}$$

- After reaching the final steady value of current I, energy input into the inductor stops $\rightarrow$ energy is stored in the magnetic field of the inductor
- When the current decreases from I to zero, the inductor acts as a source that supplies a total energy of $\frac{LI^2}{2}$ to the circuit

NOTE Resistor $\rightarrow$ energy flows into it whenever current passes through it (steady or time-varying)
 $\rightarrow$ energy is dissipated as heat

X

Inductor $\rightarrow$ energy flows into it only when current passing through it increases
 $\rightarrow$ energy flows out of it when current passing through it decreases

→ for a steady current in an inductor,
there is no energy flow in or out of the inductor

Magnetic energy density

- Energy in an inductor is stored in its magnetic field
- Let us consider an ideal toroidal solenoid
 - magnetic field is completely confined to a finite space within the toroid core
 - magnetic field is uniform across the toroid cross-section
 - total volume enclosed by the toroid is approximately

$$V = 2\pi r \cdot A$$

↓

$$\text{Total stored energy } U = \frac{1}{2} L I^2 = \frac{1}{2} \frac{\mu_0 A N^2}{2\pi r} \cdot I^2$$

is localized in the volume V

↓

$$\begin{aligned} \text{Energy density } u &\equiv \frac{U}{V} = \frac{1}{2} \frac{\mu_0 A N^2}{2\pi r} I^2 \cdot \frac{1}{2\pi r A} \\ &= \frac{1}{2} \mu_0 \frac{N^2 I^2}{(2\pi r)^2} \end{aligned}$$

- If we realize that the magnetic field of the toroid is

$$B = \mu_0 \frac{NI}{2\pi r}, \text{ we can rearrange to obtain } \frac{B}{\mu_0} = \frac{NI}{2\pi r}$$

→ Substituting this into the definition of magnetic energy density gives

$$\boxed{u = \frac{1}{2} \mu_0 \left(\frac{B^2}{\mu_0^2} \right) = \frac{1}{2} \frac{B^2}{\mu_0}} \quad \text{— magnetic energy density in vacuum}$$

- If the material inside the toroid has a constant magnetic permeability μ , the energy density is

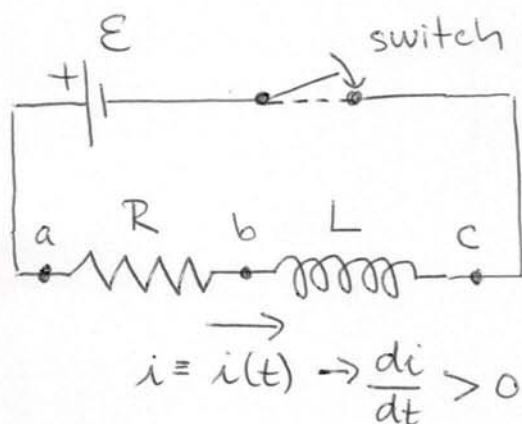
$$\boxed{u = \frac{1}{2} \frac{B^2}{\mu}}$$

→ this expression is valid generally for any magnetic field configuration in a material with constant permeability

Section 30.4: The R-L Circuit

- An electric circuit that includes a resistor and an inductor

a) Current growth in an R-L circuit



- switch is closed at time $t=0 \Rightarrow$ current starts flowing through the circuit



induced emf appears across inductor that opposes the current growth

- potential difference across the resistor: $V_{ab} = iR > 0$
- potential difference across the inductor: $V_{bc} = L \frac{di}{dt} > 0$

Kirchhoff's loop rule for the circuit in the counter-clockwise direction:

$$+\varepsilon - iR - L \frac{di}{dt} = 0$$



Rate of current increase: $\frac{di}{dt} = \frac{\varepsilon}{L} - \frac{R}{L} \cdot i$

- When the switch is first closed, $i=0$

→ initial rate of current change $\underline{\underline{\left(\frac{di}{dt}\right)_{\text{initial}} = \frac{\mathcal{E}}{L}}}$

- After a long time, current approaches a steady value I

$$\Rightarrow \left(\frac{di}{dt}\right)_{\text{final}} = 0 \Rightarrow \frac{\mathcal{E}}{L} - \frac{R}{L} I = 0 \Rightarrow \underline{\underline{I = \frac{\mathcal{E}}{R}}}$$

- - final current does not depend on the inductance, only on resistance

Current evolution with time

- Let us rearrange the Kirchhoff's loop rule $\frac{di}{dt} = \frac{\mathcal{E}}{L} - \frac{R}{L} i$

as $\frac{di}{i - \frac{\mathcal{E}}{R}} = -\frac{R}{L} dt$ and integrate both sides:

$$\int_0^i \frac{di'}{i' - \frac{\mathcal{E}}{R}} = - \int_0^t \frac{R}{L} dt' \Rightarrow \left[\ln\left(i' - \frac{\mathcal{E}}{R}\right) \right]_0^i = -\frac{R}{L} [t']_0^t$$

$$\ln\left(\frac{i - \mathcal{E}/R}{-\mathcal{E}/R}\right) = -\frac{R}{L} t$$

- If we take exponentials of both sides, we obtain

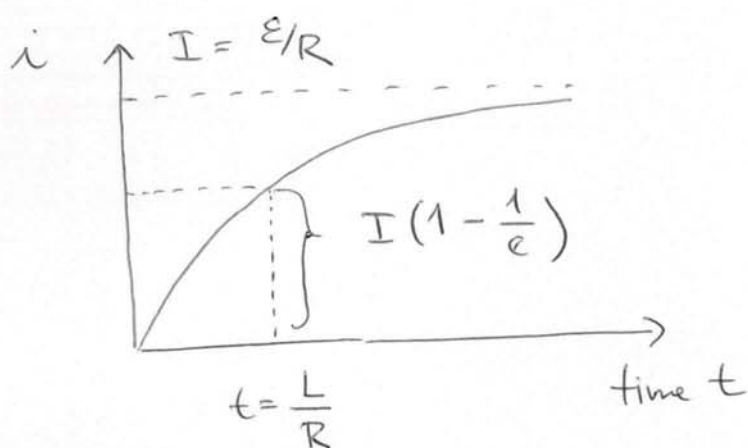
$$\frac{i - \mathcal{E}/R}{-\mathcal{E}/R} = e^{-\frac{R}{L} t} \text{ which can be rearranged as } \rightarrow$$

$$i = \frac{\mathcal{E}}{R} \left(1 - e^{-\frac{R}{L}t} \right)$$

time-varying current in R-L circuit when the current goes from zero to a finite steady value $\frac{\mathcal{E}}{R}$.

- Quantity $\tau = \frac{L}{R}$ defines the time constant of R-L circuit $\rightarrow$ it characterizes how quickly the current builds up towards its final value

NOTE Remember time constant for R-C circuit $\tau = R \cdot C$ that describes dynamics of capacitor charging



$\rightarrow$ for $t \rightarrow \infty$, inductor behaves as a simple conducting element with zero resistance

Energy considerations SKIP

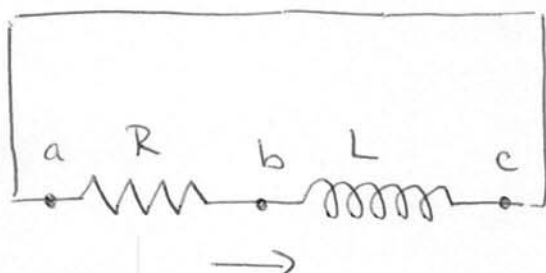
- power delivered to the circuit: $P_{in} = \mathcal{E} \cdot i$
- power dissipated in the resistor: $P_R = i^2 R$
- power stored in the inductor: $P_L = \frac{d}{dt} \left(\frac{1}{2} Li^2 \right) = Li \frac{di}{dt}$

Energy balance requires $P_{in} = P_R + P_L$

$$\Downarrow$$

$$\boxed{\mathcal{E}i = i^2R + Li \frac{di}{dt}}$$

b) Current decay in an R-L circuit



$$i \equiv i(t) \rightarrow \frac{di}{dt} < 0$$

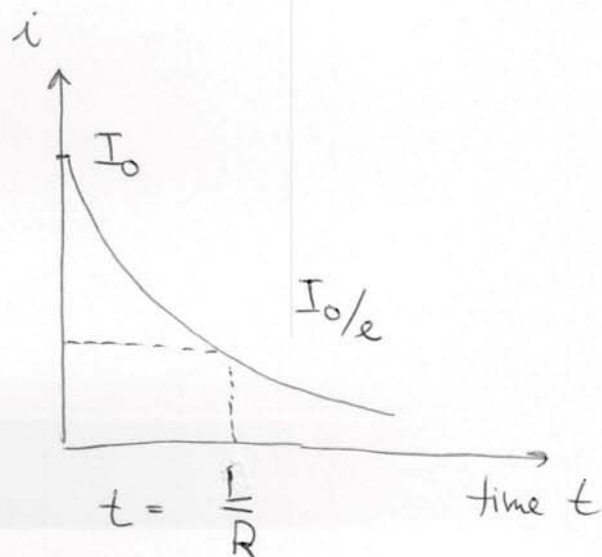
Kirchhoff's loop rule requires

$$iR + L \frac{di}{dt} = 0$$

$$\int_{I_0}^i \frac{di'}{i'} = -\frac{R}{L} \int_0^t dt' \quad \Downarrow \Rightarrow \quad \ln \frac{i}{I_0} = -\frac{R}{L} t$$

$$\Downarrow$$

$$\boxed{i = I_0 e^{-\frac{R}{L} t}}$$



Energy balance:

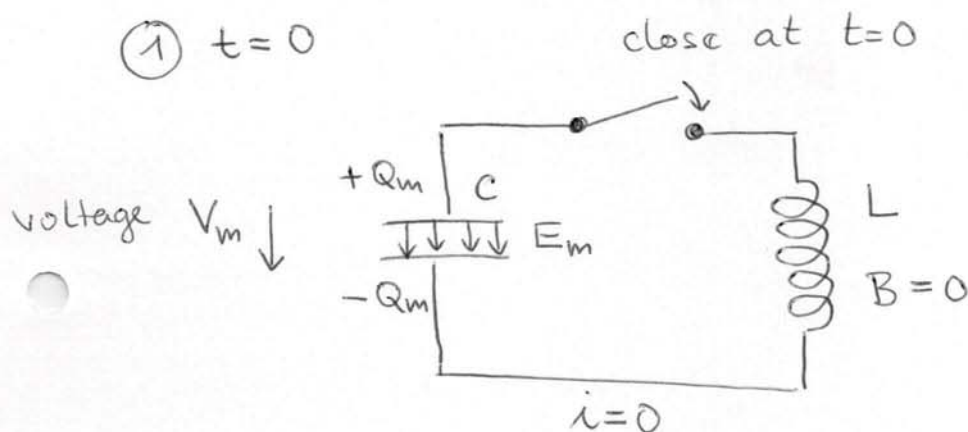
$$0 = i^2R + Li \frac{di}{dt}$$

$\Downarrow$
energy stored in the inductor
is dissipated in the resistor

Section 30.5: The L-C Circuit

- A circuit containing an inductor and a capacitor shows an entirely new mode of behavior
 → charge and current oscillate in time

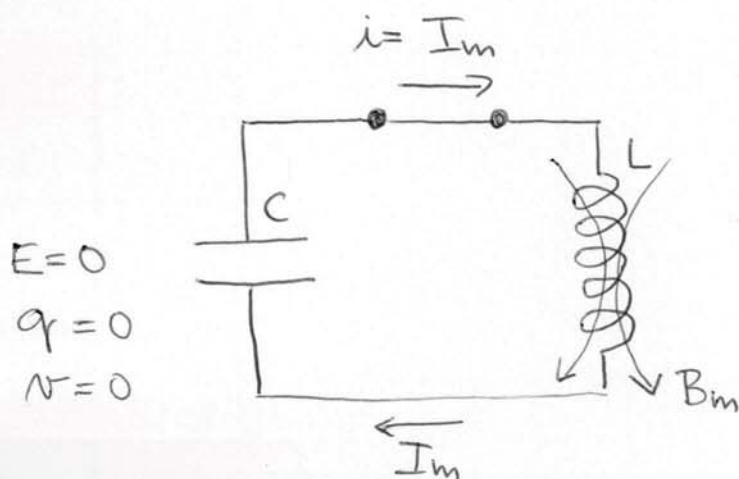
① $t = 0$



- initial current is zero → no magnetic field
 → all energy stored in the electric field of the capacitor

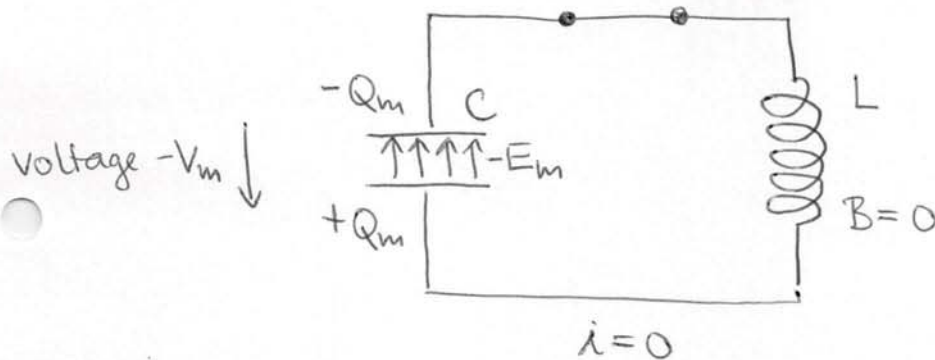
↓ capacitor discharges, current builds up to a maximum value I_m

② $t = \frac{T}{4}$



current starts decaying,
induced emf tries to keep it going,
capacitor charges with opposite
polarity

③ $t = \frac{T}{2}$



At time $t = \frac{T}{2}$, capacitor
is fully charged with
reversed polarity
→ maximal charge $-Q_m$,
voltage $-V_m$, and
electric field $-E_m$

- current is again zero → no magnetic field in
the inductor → all energy stored in the electric field
of the capacitor

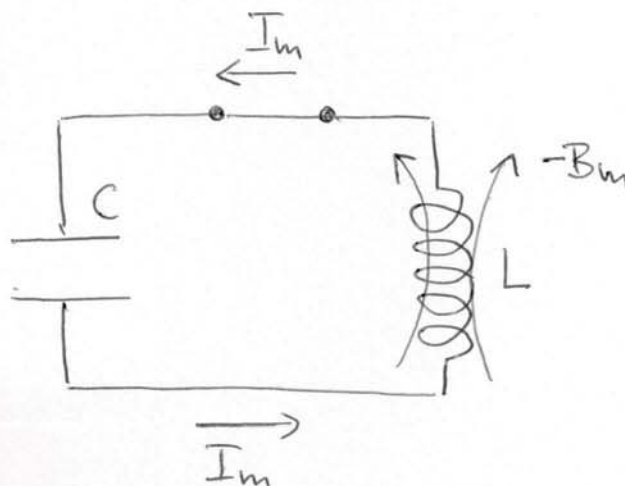
capacitor discharges, current with
reversed polarity builds up to maximal
value $-I_m$

④ $t = \frac{3T}{4}$

$E=0$

$q=0$

$N=0$



At time $t = \frac{3T}{4}$,
capacitor is again
fully discharged
→ zero voltage and
electric field

- maximal reversed current $-I_m$ → all energy stored in
the magnetic field of the inductor

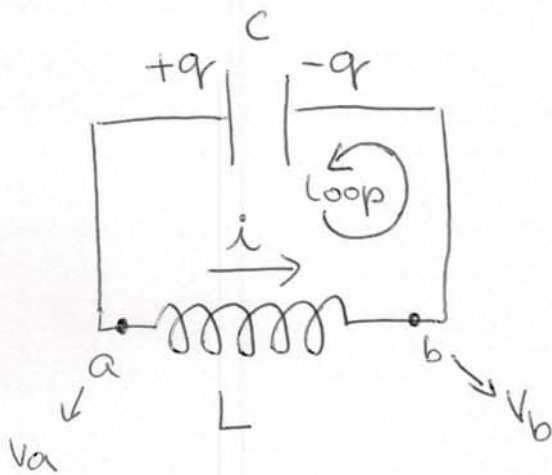
- Full oscillation cycle is completed by full charging of the capacitor to the original value of charge at time $t=0 \rightarrow$ the whole process repeats again, in the absence of dissipation indefinitely



electrical oscillation

- During oscillations, energy transfers between capacitor's electric field and inductor's magnetic field
 - $\rightarrow$ the total energy remains constant
 - $\rightarrow$ analogy to the transfer between potential and kinetic energy during mechanical oscillations

Electrical oscillations in an LC circuit



Consider a circuit containing an (ideal) inductor and capacitor



No energy dissipation in the circuit



Energy oscillates between the capacitor and inductor

- Let's assume the charge is decreasing $\rightarrow$ Kirchhoff's loop

rule gives $\frac{q}{C} - L \frac{di}{dt} = 0$ (current grows with time)

- Current i with positive direction indicated in the figure is linked to the time rate of change of the charge q on the left capacitor plate by

$$\boxed{i = - \frac{dq}{dt}} \rightarrow \text{decreasing charge on the left plate causes positive current}$$



Kirchhoff's loop rule: $\frac{q}{C} - L \frac{d}{dt} \left(-\frac{dq}{dt} \right) = 0$

Dynamics of the capacitor charge

$$\boxed{\frac{d^2 q}{dt^2} + \frac{1}{LC} \cdot q = 0}$$

⇕ analogy with mechanical oscillations

Dynamics of the oscillator displacement

$$\boxed{\frac{d^2 x}{dt^2} + \frac{k}{m} x = 0}$$

Electrical oscillation		Mechanical oscillation	
charge	q	↔	x (displacement)
current	$i = \frac{dq}{dt}$	↔	velocity $v = \frac{dx}{dt}$
inductance	L	↔	mass m
capacitance	C	↔	reciprocal force constant $\frac{1}{k}$

- Solutions of differential equation for charge q has the form

$$q(t) = Q \cos(\omega t + \varphi) \quad \leftarrow \text{harmonic oscillation}$$

with frequency

$$\omega = \sqrt{\frac{1}{LC}}$$

↓

Current

$$i(t) = -\frac{dq}{dt} = \omega Q \sin(\omega t + \varphi)$$

- Constants $\underline{Q}$ and $\underline{\varphi}$ are determined by initial conditions

$$\textcircled{\text{EG}} \quad \left. \begin{array}{l} t = 0 \rightarrow q \text{ is maximal} \\ i \text{ is zero} \end{array} \right\} \Rightarrow \begin{array}{l} Q = Q_{\max} \\ \varphi = 0 \end{array}$$

Energy in an L-C circuit

- Inductor energy $U_L = \frac{1}{2} L i^2 = \frac{1}{2} L \omega^2 Q_{\max}^2 \sin^2\left(\frac{t}{\sqrt{LC}}\right)$

$$= \frac{1}{2} L \frac{1}{LC} Q_{\max}^2 \sin^2\left(\frac{t}{\sqrt{LC}}\right) =$$

$$= \frac{Q_{\max}^2}{2C} \sin^2\left(\frac{t}{\sqrt{LC}}\right)$$

- Capacitor energy $U_C = \frac{q^2}{2C} = \frac{Q_{\max}^2}{2C} \cos^2\left(\frac{t}{\sqrt{LC}}\right)$

• Total energy $U_T = U_L + U_C =$

$$= \frac{Q_{\max}^2}{2C} \sin^2\left(\frac{t}{\sqrt{LC}}\right) + \frac{Q_{\max}^2}{2C} \cos^2\left(\frac{t}{\sqrt{LC}}\right) =$$

$$= \frac{Q_{\max}^2}{2C} = \text{const}$$

Ex 30.9: An oscillating circuit

A 300 V dc power supply is used to charge a $25 \mu\text{F}$ capacitor. After the capacitor is fully charged, it is disconnected from the power supply and connected across a 10 mH inductor.

a) Find frequency and period of circuit oscillation

$$\omega = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{10 \cdot 10^{-3} \cdot 25 \cdot 10^{-6}}} = \underline{\underline{2000 \frac{\text{rad}}{\text{s}}}}$$

$$\omega = \frac{2\pi}{T} \rightarrow T = \frac{2\pi}{\omega} = \underline{\underline{3,14 \text{ ms}}}$$

b) Find the capacitor charge and circuit current $1,2 \text{ ms}$ after the inductor and capacitor are connected

$$q = Q_{\max} \cos \omega t$$

$$i = -\frac{dq}{dt} = \omega Q_{\max} \sin \omega t$$

$$Q_{\max} = C \cdot V = 25 \cdot 10^{-6} \mu\text{F} \cdot 300 \text{ V} = \underline{\underline{7,5 \cdot 10^{-3} \text{ C}}}$$

$$q = 7,5 \cdot 10^{-3} \cdot \cos(2000 \cdot 1,2 \cdot 10^{-3}) = - \underline{\underline{5,5 \cdot 10^{-3} \text{ C}}}$$

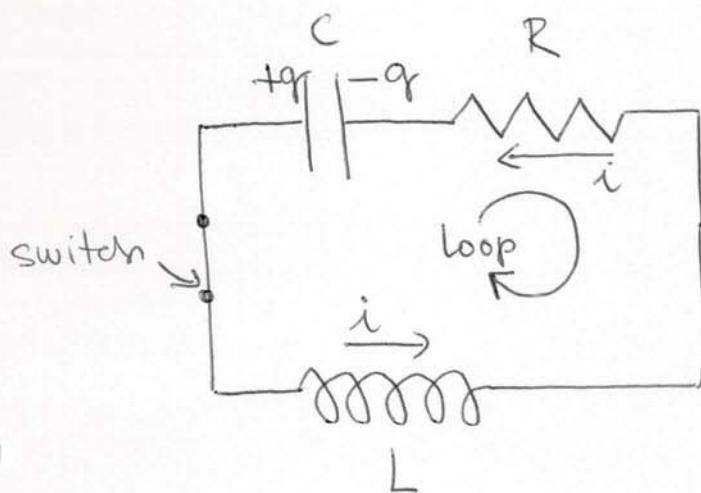
$$i = 2000 \cdot 7,5 \cdot 10^{-3} \cdot \sin(2000 \cdot 1,2 \cdot 10^{-3}) = \underline{\underline{10,1 \text{ A}}}$$

Section 30.6: The L-R-C Circuit - Series

- Including resistance R in an oscillating LC circuit causes energy dissipation $\rightarrow$ total stored energy decays towards zero for long times



- For small resistance, circuit displays damped harmonic oscillation
- For increasing resistance, circuit behavior loses oscillatory character



At time $t=0$, capacitor is fully charged to charge $Q = E \cdot C$



closing the switch causes capacitor discharging

Kirchhoff's loop rule:

$$-\frac{q}{C} + iR + L \frac{di}{dt} = 0$$

Current $i = -\frac{dq}{dt} \rightarrow$ rate of change of the charge on the left capacitor plate



$$-L \frac{d^2q}{dt^2} - R \frac{dq}{dt} - \frac{q}{C} = 0 \rightarrow \boxed{\frac{d^2q}{dt^2} + \frac{R}{L} \frac{dq}{dt} + \frac{1}{LC} q = 0}$$

- second-order differential equation for charge q

- solutions have the general form $q(t) = A \cdot e^{kt}$
 $\downarrow$ insert into differential equation

$$A k^2 e^{kt} + \frac{R}{L} A k e^{kt} + \frac{A}{LC} e^{kt} = 0$$

$$A e^{kt} \left(k^2 + \frac{R}{L} k + \frac{1}{LC} \right) = 0 \rightarrow \text{valid for all times}$$

$\Downarrow$

$$k^2 + \frac{R}{L} k + \frac{1}{LC} = 0$$

Two solutions of the quadratic equation

$$k_{1,2} = -\frac{R}{2L} \pm \frac{1}{2} \sqrt{\frac{R^2}{L^2} - \frac{4}{LC}}$$

General solution:

$$q(t) = A e^{k_1 t} + B e^{k_2 t} =$$

$$= e^{-\frac{R}{2L} t} \cdot \left\{ A e^{+\frac{1}{2} \sqrt{\frac{R^2}{L^2} - \frac{4}{LC}} t} + B e^{-\frac{1}{2} \sqrt{\frac{R^2}{L^2} - \frac{4}{LC}} t} \right\}$$

- There are three possible regimes of operation depending on the values of R , L , and C

$$a) \frac{R^2}{L^2} = \frac{4}{LC} \rightarrow \underline{q(t) = (A+B)e^{-\frac{R}{2L}t}}$$

- critically damped case with exponential decay of the charge

b) $\frac{R^2}{L^2} > \frac{4}{LC} \rightarrow q(t)$ given by the sum of two decreasing exponentials with different time constants
- overdamped case

c) $\frac{R^2}{L^2} < \frac{4}{LC} \rightarrow$ under damped case

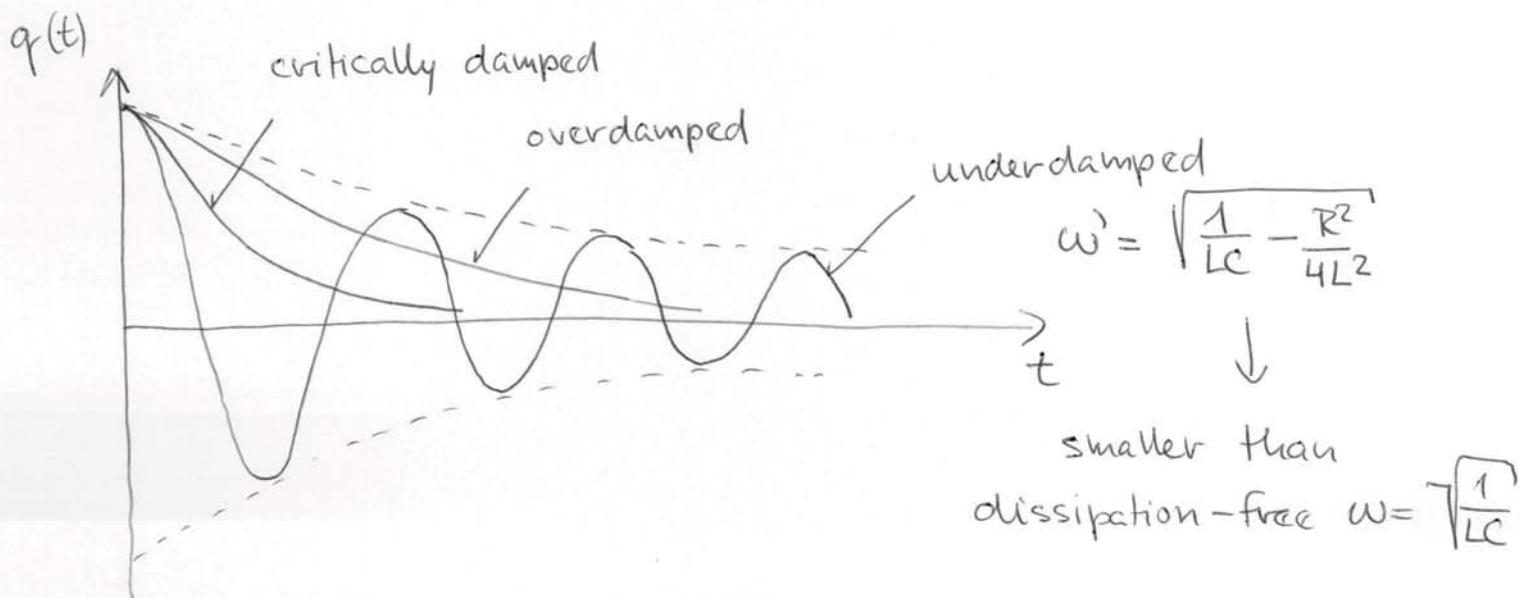
$$q(t) = e^{-\frac{R}{2L}t} \left\{ A e^{+\frac{i}{2}\sqrt{\frac{4}{LC} - \frac{R^2}{L^2}}t} + B e^{-\frac{i}{2}\sqrt{\frac{4}{LC} - \frac{R^2}{L^2}}t} \right\}$$

exponential damping

oscillatory motion

$$K \cdot \cos \left[\frac{1}{2}\sqrt{\frac{4}{LC} - \frac{R^2}{L^2}}t + \phi \right]$$

$$\underline{q(t) = K e^{-\frac{R}{2L}t} \cdot \cos \left[\sqrt{\frac{1}{LC} - \frac{R^2}{4L^2}}t + \phi \right]}$$



Ex 30.11: Underdamped L-R-C series circuit

What resistance R is required to give an L-R-C circuit a frequency that is one-half the undamped frequency?

Underdamped frequency $\omega' = \sqrt{\frac{1}{LC} - \frac{R^2}{4L^2}}$

Undamped frequency $\omega = \sqrt{\frac{1}{LC}}$

$$\omega' = \frac{1}{2}\omega \rightarrow \sqrt{\frac{1}{LC} - \frac{R^2}{4L^2}} = \frac{1}{2} \sqrt{\frac{1}{LC}}$$

$$\frac{1}{LC} - \frac{R^2}{4L^2} = \frac{1}{4LC}$$

$$\frac{3}{4LC} = \frac{R^2}{4L^2} \Rightarrow \underline{\underline{R = \sqrt{\frac{3L}{C}}}}$$