

Bifurcation analyses and hardware experiments for bursting dynamics in non-autonomous memristive FitzHugh-Nagumo circuit

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In this paper, a non-autonomous memristive FitzHugh-Nagumo (FHN) circuit is constructed using a second-order memristive diode bridge with LC network. For convenience of circuit implementation, an AC voltage source is adopted to substitute the original AC current stimulus. Stimulated by the slowly varying AC voltage source, the number, locations and stabilities of the equilibrium points slowly evolve with the time, which are thus indicated as the AC equilibrium points. Different sequences of fold and/or Hopf bifurcations are encountered in a full period of time series evolutions, leading to various kinds of chaotic or periodic bursting activities. To figure out the related bifurcation mechanisms, the fold and Hopf bifurcation sets are mathematically formulated to locate the critical bifurcation points. On this basis, the transitions between the resting and repetitive spiking states are clearly illustrated by the time series of the AC equilibrium points and state variables, from which Hopf/subHopf, Hopf/Hopf, and Hopf/fold bursting oscillations are identified in the specified parameter regions. Finally, based on a fabricated hardware circuit, the experimental measurements are executed. The results verify that the presented memristive FHN circuit indeed exhibits complex bursting activities, which enriches the family of memristor-based FHN circuits with bursting dynamics.

FitzHugh-Nagumo circuit, memristive diode bridge, bursting, AC equilibrium point

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1 Introduction

In the biological nervous systems, bursting activities play important roles in information encoding and transmission, memory formation [1,2], and occurrence of neuronal disease [3]. Numerous mathematical neuronal models [4–8] have been established to promote theoretical investigations of the electrical activities in single and coupled neurons. The two-dimensional FitzHugh-Nagumo (FHN) model [9,10] is simplified from the four-dimensional Hodgkin-Huxley model for geometrical explanations of neuronal excitability and spiking under excitation. Many researches have been

done on the significant and complex dynamical aspects of the FHN model. For example, Alidousti and Ghaziani [2] discussed various kinds of bursting activities in autonomous fractional-order FHN model. Baltanás and Casado [11] reported spontaneous bursting activities in FHN model subjected to quasi-monochromatic noise. Abbasian et al. [12] analyzed symmetric bursting activities in autonomous generalized FHN model. Besides, some specific dynamical behaviors, including chaos and bifurcation [13,14], noise or time delay influence [15–17], and synchronization characteristics [18,19], were also investigated by theoretical analyses, numerical simulations, and hardware experiments.

Due to the synaptic plasticity and non-volatile memory, memristor is becoming an excellent candidate for hardware

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implementations of neuromorphic circuits. At device levels, memristors have been proved to be able to emulate synaptic functions by storing analog synaptic weight or implementing synaptic learning rules [20]. Thus, memristors were used as essential building blocks for memristor crossbar-based neuromorphic computing or learning systems [20–22]. Besides, the effects of electromagnetic radiation [23–25] and coupling relationship between neuronal cells [26–29] can also be depicted by the memristive synapses. Inspired by these, some memristor-based FHN models have been proposed recently. Zhang et al. [24] used memristors with quadratic memductance to denote the nonlinear autapse or synapse current applied on isolated FHN model or occurred between FHN models in the chain network. Mvogo et al. [30] studied spatiotemporal formation of patterns in a diffusive FHN network where the modulation of magnetic flux on membrane potential was realized by memristor coupling. Ma et al. [31] investigated the effect of electric field in isolate neurons by introducing a new field variable E and distinct mode transitions of the electric activities were observed. Furthermore, by using a memristor synapse with non-smooth piecewise-linear memductance to realize the nonlinear membrane voltage function of the FHN model, Zhang and Liao [32,33] reported a chaotic non-autonomous FHN circuit and investigated the synchronization performances of memristive FHN circuits coupled by memristor synapse. Thereafter, Bao et al. [34] used an ideal memristor with hyperbolic tangent memductance to replace the memristor of refs. [32,33] and hidden extreme multistability was revealed and investigated through dimensionality reduction analyses. However, bursting activities and the related bifurcation mechanisms were rarely discussed in non-autonomous memristive circuits [35].

The bursting oscillation refers to oscillation transitions between the resting and repetitive spiking states caused by the modulation between the fast-scale and slow-scale variables. It has been found in a large variety of neuronal models [2,11,12,36–38] and nonlinear dynamical systems [39–43]. From the perspective of dynamical systems, the possible bifurcations and their role in the computational properties of bursting activities can be illustrated using geometrical methods [36]. For the autonomous systems, the fast-slow dissection method [40,43] can readily be applied for classifications of bursting activities. While for non-autonomous systems with a slowly varying periodic excitation, the exciting term is commonly regarded as a standalone control variable, then bifurcation mechanisms of the revealed bursting activities are explored using the so-called transformed phase portraits [39,41,42].

In this paper, aiming to perform quantitative analyses and experimental verifications of curious bursting activities in the non-autonomous memristive FHN model, a new neural model is constructed using the AC voltage source and sec-

ond-order memristive diode bridge emulator. Driven by the slowly varying external force, different bursting activities are uncovered. Specially, the system trajectory passes a bifurcation point twice in a full period of the busting oscillation. Consequently, the transformed phase portrait overlaps together, leading to some difficulties in the bifurcation analyses. However, numerical simulations reveal that the characteristics of the AC equilibrium points provide abundant information for identifying the bifurcation mechanisms of these bursting activities. It is proved to be an efficient method for theoretical analyses of dynamics in non-autonomous systems. Moreover, since the external stimulus is implemented using an AC voltage source, the revealed bursting activities are physically realized in a hardware circuit, which has rarely been reported in previous literature and could contribute to practical engineering applications of artificial neuromorphic circuits.

2 Non-autonomous memristive FHN circuit

The non-autonomous memristive FHN circuit is achieved by employing a second-order memristive diode bridge with LC network [40] to substitute the ideal memristor possessing non-smooth memductance in ref. [32] or smooth hyperbolic tangent memductance in ref. [34]. The circuit schematics of the main circuit and second-order memristive diode bridge emulator are presented in Figure 1(a) and (b), respectively. For convenience of experimental measurements, an AC voltage source $V_S = V_m \sin(2\pi ft)$, connected in series with a negative resistor R_1 , is used to implement the external AC current excitation of the conventional FHN model.

2.1 Mathematical model

The voltage-current relation of the second-order memristive diode bridge emulator is given as [40]

$$i_M = 2I_S e^{-\rho v_0} \sinh(\rho v_C), \quad (1a)$$

$$\begin{cases} C_0 dv_0 / dt = 2I_S [e^{-\rho v_0} \cosh(\rho v_C) - 1] - i_0, \\ L_0 di_0 / dt = v_0. \end{cases} \quad (1b)$$

Four 1N4148 diodes are adopted to achieve the memristive diode bridge, thus the device parameters are determined as $I_S = 5.84$ nA and $\rho = 10.31$ V⁻¹. Based on eq. (1), the circuit equations of the non-autonomous memristive FHN circuit are established as

$$\begin{aligned} C dv_C / dt &= (V_S - v_C) / R_1 - i_L - 2I_S e^{-\rho v_0} \sinh(\rho v_C), \\ L di_L / dt &= v_C + E - i_L R, \\ C_0 dv_0 / dt &= 2I_S [e^{-\rho v_0} \cosh(\rho v_C) - 1] - i_0, \\ L_0 di_0 / dt &= v_0. \end{aligned} \quad (2)$$

Denote

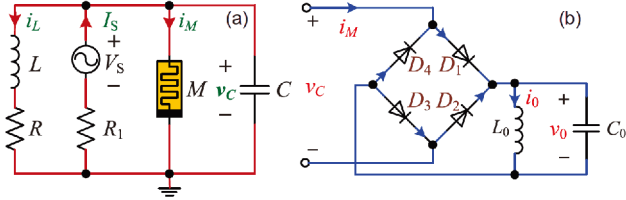


Figure 1 (Color online) Non-autonomous memristive FHN circuit. (a) Main circuit; (b) second-order memristive diode bridge emulator.

$$\begin{aligned} x &= \rho v_C, y = \rho R i_L, z = \rho v_0, u = \rho R i_0, \\ w &= A \sin(F\tau), t = RC\tau, \\ A &= \rho V_m, F = 2\pi fRC, D = 2\rho R I_s, \\ k_r &= R/R_1, k_c = C/C_0, l = R^2 C/L, l_0 = R^2 C/L_0. \end{aligned} \quad (3)$$

The normalized model of eq. (2) is formulated as follows:

$$\begin{aligned} \dot{x} &= k_r(w - x) - y - D e^{-z} \sinh x, \\ \dot{y} &= l(x - y), \\ \dot{z} &= k_c(D e^{-z} \cosh x - D - u), \\ \dot{u} &= l_0 z. \end{aligned} \quad (4)$$

In eq. (4), there are the stimulus-related parameters A and F in the slowly varying variable w , diode-related parameter D , and other circuit-element-related control parameters k_r , k_c , l , and l_0 . When the circuit element parameters of Figure 1 are assigned as $V_m=1$ V, $f=0.32$ kHz, $C=10$ nF, $C_0=3.3$ nF, $L=15$ mH, $L_0=5$ mH, $R=1$ k Ω , and $R_1=-1.25$ k Ω , these control parameters are calculated as

$$\begin{aligned} A &= 10.31, F = 0.02, D = 1.204 \times 10^{-4}, \\ k_r &= -0.80, k_c = 3.03, l = 0.667, l_0 = 2. \end{aligned} \quad (5)$$

In this case, a chaotic bursting pattern with $LE_1=0.0124$, $LE_2=-0.0134$, $LE_3=-0.4901$, and $LE_4=-12.8254$ is revealed. The phase portrait and time series are depicted in Figure 2(a) and (b), respectively. Obviously, the external forcing frequency F is much smaller than the natural frequency of system (4). So the slowly varying AC voltage stimulus modulates the fast variables and makes them switch between the resting and spiking states, leading to the occurrence of bursting oscillations.

Since the bursting activities caused by different bifurcation mechanisms are what we mainly concerned, k_r (i.e. the negative resistor R_1) and l (i.e. the inductor L) are selected as main control variables in later sections. These two parameters affect the bifurcation sequences encountered in a full period of time series evolutions, resulting in various kinds of chaotic or periodic bursting activities.

2.2 AC equilibrium point and stability

In non-autonomous systems, the equilibrium points are evolved with the variations of the external stimulus, which are thus indicated as the AC equilibrium points [44,45].

Letting $\dot{x} = \dot{y} = \dot{z} = \dot{u} = 0$, the AC equilibrium point of system (4) is expressed as

$$E_{AC} = (x_e, x_e, 0, D \cosh x_e - D), \quad (6)$$

where x_e is solved from the following equation:

$$F_0 = (k_r + 1)x_e + D \sinh x_e - k_r w = 0. \quad (7)$$

The characteristic equation for evaluating the stability of E_{AC} is derived from the Jacobian matrix at E_{AC} as

$$\det(\mathbf{1}\lambda - \mathbf{J}) = \lambda^4 + a_3\lambda^3 + a_2\lambda^2 + a_1\lambda + a_0 = 0, \quad (8)$$

where

$$\begin{aligned} a_3 &= (k_c + 1)D \cosh x_e + k_r + l, \\ a_2 &= (k_r k_c + k_c l + l)D \cosh x_e + (k_r + 1)l + D^2 k_c + k_c l_0, \\ a_1 &= k_c(k_r l + l + l_0)D \cosh x_e + D^2 k_c l + k_c l_0(k_r + l), \\ a_0 &= k_c l l_0(D \cosh x_e + k_r + 1). \end{aligned} \quad (9)$$

Based on eqs. (7) and (8), the characteristics of E_{AC} can be readily evaluated. In addition, for the special case of $w=0$, system (4) has the DC equilibrium point characterized by

$$E_{DC} = \{(x_e, x_e, 0, D \cosh x_e - D) | (k_r + 1)x_e + D \sinh x_e = 0\}. \quad (10)$$

When $k_r \geq -1$, there is only one DC equilibrium point $E_{DC0} = (0, 0, 0, 0)$; while for $k_r < -1$, there are three DC equilibrium points indicated as E_{DC0} and $E_{DC\pm}$.

Figure 3(a) and (b) illustrate the trajectories of E_{AC} in the w - x_e plane for $l=0.667$ and $k_r=-0.8/-1.2$, respectively. The DC equilibrium points are marked using black dots. It can be seen that the number, locations, and stabilities of the AC equilibrium points vary along with the external stimulus w , from which the fold and Hopf bifurcation scenarios are readily figured out. And above all, the system orbit naturally evolves along the time-domain trajectory of the AC equilibrium point, which can provide distinct and abundant information for bifurcation analysis of the revealed bursting activities.

3 The fold and Hopf bifurcation sets

To locate the critical points of the fold and Hopf bifurcations in the time-domain trajectories of the AC equilibrium points, the formulations of the fold and Hopf bifurcation sets are mathematically deduced. On this basis, the mechanisms responsible for the bursting oscillations can be readily figured out.

3.1 The fold bifurcation set

Differentiate F_0 with respect to x_e . It yields

$$F_1 = D \cosh x_e + k_r + 1 = 0, \quad (11)$$

which gives two zero points, written as

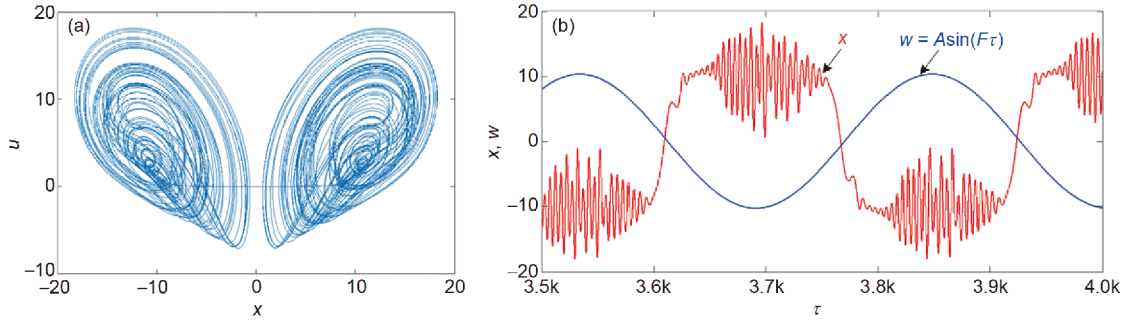


Figure 2 (Color online) Typical chaotic bursting pattern with $k_r = -0.8$ and $l = 0.667$. (a) Phase portrait in the x - u plane; (b) time series of the fast state variable x and slowly varying stimulus w .

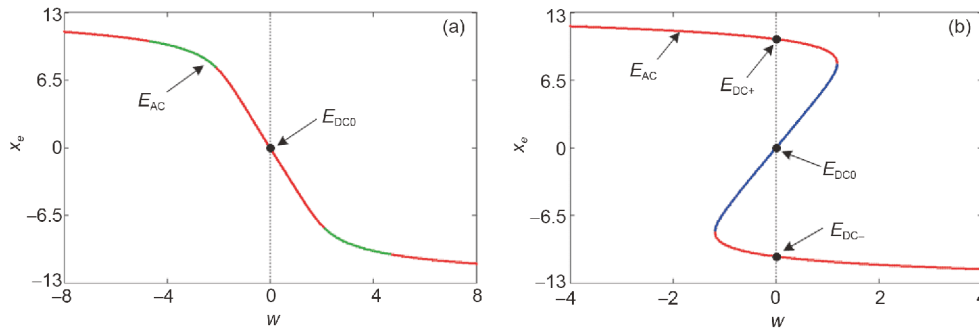


Figure 3 (Color online) AC and DC equilibrium points in the w - x_e plane, where the red, blue, and green colors represent unstable index-2 saddle-focus, unstable index-1 saddle-focus, and stable node-focus, respectively. (a) $k_r = -0.8$; (b) $k_r = -1.2$.

$$x_a = \operatorname{arccosh} \frac{-k_r - 1}{D}, \quad x_b = -\operatorname{arccosh} \frac{-k_r - 1}{D}. \quad (12)$$

When $k_r \leq -1 - D$, x_a and x_b have real roots, then F_0 reaches its extreme values, leading to the generation of the fold bifurcation (FB).

From the mathematical point of view, the FB refers to one zero eigenvalue [40,41], i.e. $a_0 = 0$ in eq. (8). Thus, the FB set can be deduced as

$$\text{FB: } \begin{cases} F_0 = (k_r + 1)x_e + D \sinh x_e - k_r w = 0, \\ a_0 = k_e l_0 (D \cosh x_e + k_r + 1) = 0, \end{cases} \quad (13)$$

and simplified as

$$\text{FB: } \pm \left[(k_r + 1) \operatorname{arccosh} \frac{-k_r - 1}{D} + D \sinh \left(\operatorname{arccosh} \frac{-k_r - 1}{D} \right) \right] - k_r w = 0. \quad (14)$$

With reference to eq. (14), the FB set can be easily calculated with fixed control parameters D and k_r .

3.2 The Hopf bifurcation set

When the eigenvalues at E_{AC} pass across the imaginary axis, the Hopf bifurcation (HB) may occur. At the critical point of HB, eq. (8) possesses a pair of pure imaginary roots $\lambda = \pm j\omega$

($\omega > 0$) and the deviation of $\operatorname{Re}(\lambda)$ with respect to the varying system parameter should not equal to zero [46,47].

Substituting $\lambda = \pm j\omega$ ($\omega > 0$) into eq. (8), we get

$$\begin{cases} a_1 - a_3 \omega^2 = 0, \\ a_0 - (a_2 - \omega^2) \omega^2 = 0. \end{cases} \quad (15)$$

Eliminating ω in eq. (15), the necessary condition for HB is expressed as

$$\text{HB: } \begin{cases} F_0 = (k_r + 1)x_e + D \sinh x_e - k_r w = 0, \\ a_1(a_3 a_2 - a_1) - a_3^2 a_0 = 0. \end{cases} \quad (16)$$

Take k_r as a varying parameter, denote the critical HB point as c_H , and differentiate eq. (8) with respect to k_r . We get the following expression:

$$\frac{d\lambda}{dk_r} = \frac{\lambda^3 + (k_e D \cosh x_e + l) \lambda^2 + (k_e D \cosh x_e + k_e l_0) \lambda + k_e l_0}{4\lambda^3 + 3a_3 \lambda^2 + 2a_2 \lambda + a_1}. \quad (17)$$

Put $\lambda(c_H) = \pm j\omega_H$ into eq. (17) and extract the real parts of the obtained equation. The sufficient condition for HB is deduced as

$$\frac{d\operatorname{Re}(\lambda)}{dc} \Big|_{\lambda(c_H) = \pm j\omega_H} = \frac{b_1 b_3 + b_2 b_4}{b_1^2 + b_2^2} \neq 0, \quad (18)$$

where

$$\begin{aligned}\omega_H &= \sqrt{a_1/a_3}, b_1 = -2a_1, b_2 = 2\omega_H(a_2 - 2\omega_0^2), \\ b_3 &= -(k_e D \cosh x_e + l)\omega_H^2 + k_e l l_0, \\ b_4 &= \omega_H(k_e l D \cosh x_e + k_e l_0 - \omega_H^2).\end{aligned}\quad (19)$$

To depict the HB set, the second implicit function of eq. (16) is first solved using MATLAB function to get the x_e values of each HB point. Then, the coordinates of the HB point and the corresponding w values can be readily obtained. These points are further verified using the sufficient condition described in eq. (18) to eliminate the points at which eq. (8) possesses pure imaginary roots but the eigenvalues do not pass the imaginary axis with the variation of k_r .

4 Bursting activities and bifurcation mechanisms

Based on the deduced FB and HB sets, as well as Izhikevich's classification scheme of bursting [36], the bursting activities of the non-autonomous memristive FHN circuit are revealed and identified. During the analyses, the parameters k_r and l are chosen as the main control variables and the initials are fixed as (0, 0, 0, 0).

4.1 Evolution of bursting activities

The variations of k_r and l lead to different combinations of FBs and HBs, and then change the exhibited dynamics of the non-autonomous memristive FHN circuit. Take $l=0.667$ and 3 as two examples. The FB and HB sets are depicted in the w - k_r plane, as shown in Figure 4. It can be found that for $l=0.667$, the HBs occur in the range of $-1.183 \leq k_r \leq -0.628$ and FBs emerge within $-2 \leq k_r < -1$; while for $l=3$, the HBs are revealed in the region of $-2 \leq k_r \leq -0.997$ and the FBs are found within $-2 \leq k_r < -1$. Note, to get better visual effect, only the essential regions of the bifurcation sets are depicted in Figure 4.

When $l=0.667$ is considered, four representative examples of $k_r=-0.6, -0.72, -0.9$, and -1.3 are selected to illustrate the dynamical evolutions. The time series of x overlapped with the x_e trajectories of E_{AC} are presented in Figure 5, in which the stabilities of E_{AC} are indicated using different colors as defined in Figure 3. At $k_r=-0.6$, there are two HB points located beyond the amplitude of the system orbit and the AC equilibrium points always behave as stable node-foci, as shown in Figure 5(a). Accordingly, the system orbit settles down to the resting state and a periodic spiking pattern is uncovered. Thereafter, the dynamics of system (4) gradually enters into the bursting activity at around $k_r=-0.672$. As depicted in Figure 5(b), a periodic bursting with negative largest Lyapunov exponent is observed at $k_r=-0.72$. The system orbit passes each of the four HB points twice in one

bursting cycle, which is different from those reported in refs. [36,40,41]. When k_r is further decreased, the resting states of the bursting activities gradually reduce and finally become inconspicuous for $k_r < -0.85$. Then, the dynamics of system (4) turns into the spiking activity. Figure 5(c) illustrates a typical chaotic spiking with positive largest Lyapunov exponent uncovered at $k_r=-0.9$. Compared with Figure 5(b), the stable AC equilibrium point regions (indicated by the green lines) are too narrow to confine the large amplitude spiking states to the resting states. Thus, the spiking states are sustained in a whole cycle of the slow variable, leading to the symmetric chaotic spiking activity. At last, Figure 5(d) exhibits a chaotic spiking at $k_r=-1.3$. Under this parameter, no HBs exist, the stable AC equilibrium point regions completely disappear, and the upper and lower spiking oscillations are linked up via the FBs.

When $l=3$ is specified, the time series of x and x_e for $k_r=-0.85, -1.0, -1.5$, and -2.3 are presented in Figure 6. Similar with that illustrated in Figure 5, the dynamics of system (4) starts from periodic spiking, as shown in Figure 6(a), and gradually enters into bursting region within $-0.9 \leq k_r < -1.39$. Figure 6(b) illustrates a typical periodic bursting generated under $k_r=-1.0$. It should be noted that within $-0.9 \leq k_r \leq -1$, two HPs induce the transitions between the resting and spiking states; while for $-1 < k_r < -1.39$, two HPs and two FBs induce those oscillation state transitions. If k_r is further decreased, the regions where the AC equilibrium points behave as the stable node-foci are gradually narrowed. Consequently, the resting states, at which the orbit resides at the stable AC equilibrium points or a small amplitude oscillation, are shortened and finally vanished for $k_r < -1.39$. Thereafter, system (4) enters into the spiking region. Two representative spiking activities at $k_r=-1.5$ and -2.3 are plotted in Figure 6(c) and (d), respectively. In this parameter region, the system orbit simply oscillates around the unstable index-2 saddle-foci in the AC equilibrium trajectory, leading to the symmetric spiking activities. The existed HB and FB points do not exert any influences on the exhibited dynamics.

It is remarkable that the dynamical evolutions of Figure 6 are similar with those revealed in Figure 5, but the inner mechanisms are totally different.

4.2 Bifurcation analysis

As shown in Figure 2(a), the non-autonomous memristive FHN circuit operates in chaotic bursting pattern for $k_r=-0.8$ and $l=0.667$. To investigate the bifurcation mechanism, the time series of x , x_e (the x coordinate of the AC equilibrium point), and z are drawn in Figure 7(a), and the transformed phase portrait [41] in the w - x plane overlapped with the trajectory of E_{AC} is supplemented in Figure 7(b). With eq. (16), the four HBs are calculated as $(w, x) = (\mp 4.7283, \pm 10.2638)$ and $(w, x) = (\mp 2.1128, \pm 7.7514)$, which are further

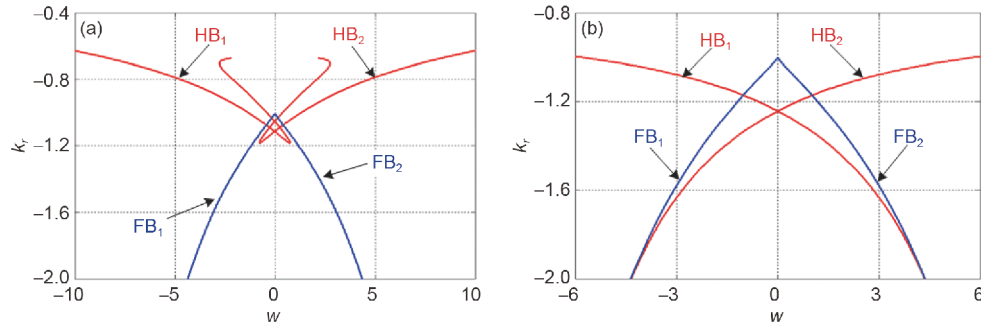


Figure 4 (Color online) HB and FB sets in the w - k_r plane. (a) $l=0.667$; (b) $l=3$.

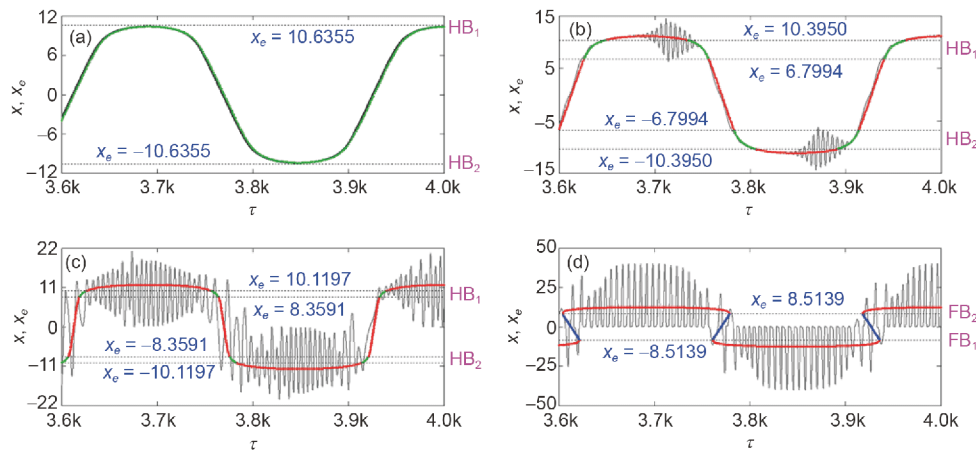


Figure 5 (Color online) Numerically simulated time series of x (the black lines) and x_e (the color lines) for $l=0.667$. (a) Periodic spiking at $k_r=-0.6$; (b) periodic bursting at $k_r=-0.72$; (c) chaotic spiking at $k_r=-0.9$; (d) chaotic spiking at $k_r=-1.3$.

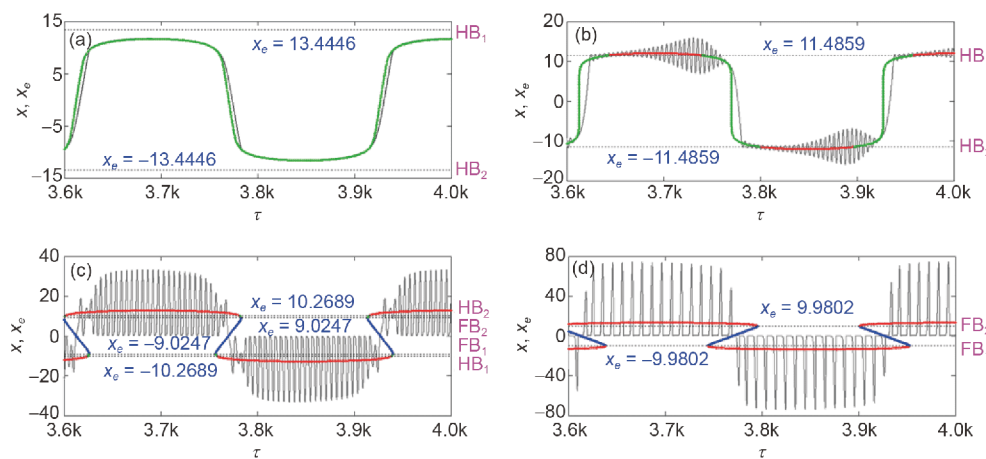


Figure 6 (Color online) Numerically simulated time series of x (the black lines) and x_e (the color lines) for $l=3$. (a) Periodic spiking at $k_r=-0.85$; (b) periodic bursting at $k_r=-1.0$; (c) periodic spiking at $k_r=-1.5$; (d) chaotic spiking at $k_r=-2.3$.

verified by eq. (18). These critical points are marked as H_1 (\hat{H}_1), H_4 (\hat{H}_4), H_2 (\hat{H}_2), and H_3 (\hat{H}_3), respectively, as seen in Figure 7.

It is well illustrated that the activity of system (4) alternates between the resting states within $H_1 \rightarrow H_4$, $\hat{H}_4 \rightarrow \hat{H}_1$ and spiking states within $A \rightarrow H_1$, $H_4 \rightarrow \hat{H}_4$, $\hat{H}_1 \rightarrow B$. The spiking

states appear via HBs at H_4 and \hat{H}_1 , and then vanish through HBs at H_1 and \hat{H}_4 . Specially, in the middle region of the resting state, i.e. $H_2 \rightarrow H_3$ ($\hat{H}_3 \rightarrow \hat{H}_2$), the AC equilibrium points turns into the unstable index-2 saddle-foci via HB at H_2 (\hat{H}_3), but only small amplitude oscillations are observed from the time series of x and z , which may be caused by the

slow passage effect [48–50]. Afterwards, through Sub-HB at $H_3(\hat{H}_2)$, the small amplitude oscillation sustains until it reaches the HB point at $H_4(\hat{H}_1)$. Consequently, this kind of bursting activity is regarded as a Hopf/SubHopf burster [36]. Since the trajectory passes each HB points twice in a full bursting period, the transitions between the resting and spiking states are hard to be distinguished from the overlapped transformed phase portrait depicted in Figure 7(b), but are clearly illustrated by the time series given in Figure 7(a).

When $k_r = -1.0/-1.2$ and $l=3$ are considered, two periodic bursting activities with negative largest Lyapunov exponents are uncovered. The time series of x , z and x_e are presented in Figure 8(a1) and (b1) to clarify the transitions of the oscillation states. The corresponding transformed phase portraits

are depicted in Figure 8(a2) and (b2), respectively. The periodic and chaotic bursting activities are briefly distinguished based on their time-domain waveforms and the calculated largest Lyapunov exponent. They can also be rigorously distinguished using 0–1 test [51,52].

For $k_r = -1.0$ and $l=3$, there are two HB points, as shown in Figure 8(a). Based on eq. (16), these two HB points are calculated to be $(w, x) = (\mp 5.8595, \pm 11.4859)$ and marked as $H_1(\hat{H}_1)$ and $H_2(\hat{H}_2)$, respectively. The upper spiking state of the bursting activity is initiated through HB at H_1 and lasts until the trajectory reaches \hat{H}_1 . Through HB at \hat{H}_1 , E_{AC} turns from the unstable index-2 saddle-focus (indicated by red lines) into the stable node-focus (indicated by green lines) and system (4) converts into the resting state. In the meantime, the trajectory moves to the lower oscillation region,

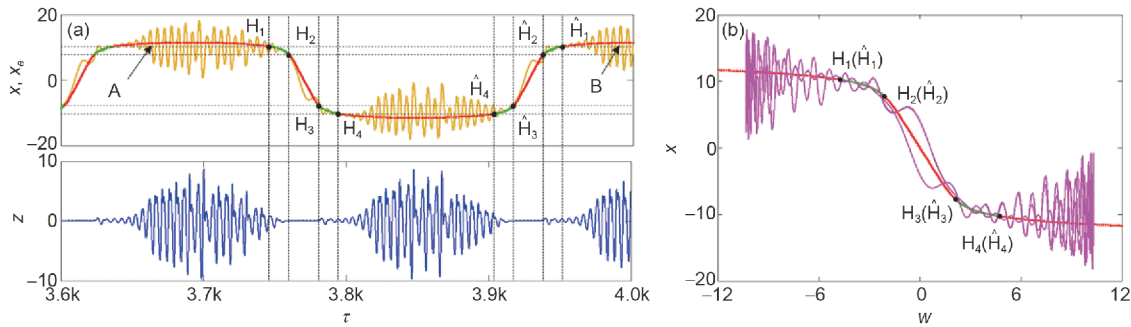


Figure 7 (Color online) Hopf/SubHopf chaotic bursting with $k_r = -0.8$ and $l=0.667$. (a) Time series of x , x_e , and z ; (b) transformed phase portrait in the w - x plane overlapped with the trajectory of E_{AC} . In E_{AC} trajectories and time series of x_e , the red and green colors represent unstable index-2 saddle-focus and stable node-focus, respectively.

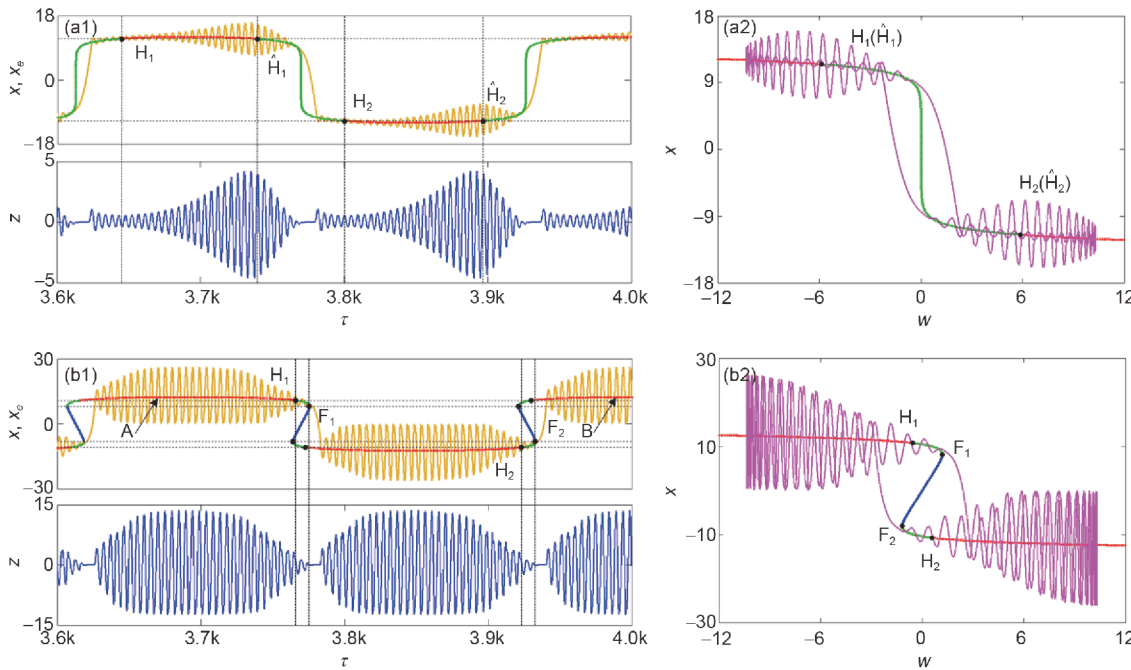


Figure 8 (Color online) Hopf/Hopf and Hopf-fold periodic bursting activities with $k_r = -1.0/-1.2$ and $l=3$, where (a1), (b1) are the time series of x , z , and x_e and (a2), (b2) are the transformed phase portraits in the w - x plane overlapped with the E_{AC} trajectory. (a) $k_r = -1.0$; (b) $k_r = -1.2$. In the E_{AC} trajectories and time series of x_e , the red, blue and green colors represent unstable index-2 saddle-focus, unstable index-1 saddle-focus, and stable node-focus, respectively.

thereby completing half of the bursting loop. Similarly, the lower spiking state emerges via HB at H_2 and fades away through HB at \hat{H}_2 . Thus, the bursting pattern of Figure 8(a) is classified as a Hopf/Hopf burster [36].

While for $k_r = -1.2$ and $l = 3$, two HBs points located at $(w, x) = (\mp 0.5795, \pm 10.7646)$ and two FB points located at $(w, x) = (\pm 1.1847, \pm 8.1084)$ are founded in the time series of x_e and E_{AC} trajectory. These points are marked as H_1 , H_2 , F_1 , and F_2 , respectively, as shown in Figure 8(b). Starting from point A, the system orbit follows along the upper branch of the E_{AC} trajectory and exhibits spiking state until it arrives at H_1 . On this point, a short-term resting state appears via HB and disappears via FB at the nearby point F_1 . Meanwhile, the

system orbit jumps to the lower branch, initiating another spiking state around the unstable lower branch E_{AC} trajectory. Similarly, the lower-branch system orbit switches from the spiking state to the resting state via HB at H_2 and reverses back to the upper branch through FB at F_2 . Accordingly, a Hopf/fold periodic bursting [36] is observed in system (4), which is somewhat similar to that reported in a memristive Wien-bridge oscillator with the same memristor emulator [40].

5 Hardware experiments

According to the circuit schematics in Figure 1, a breadboard

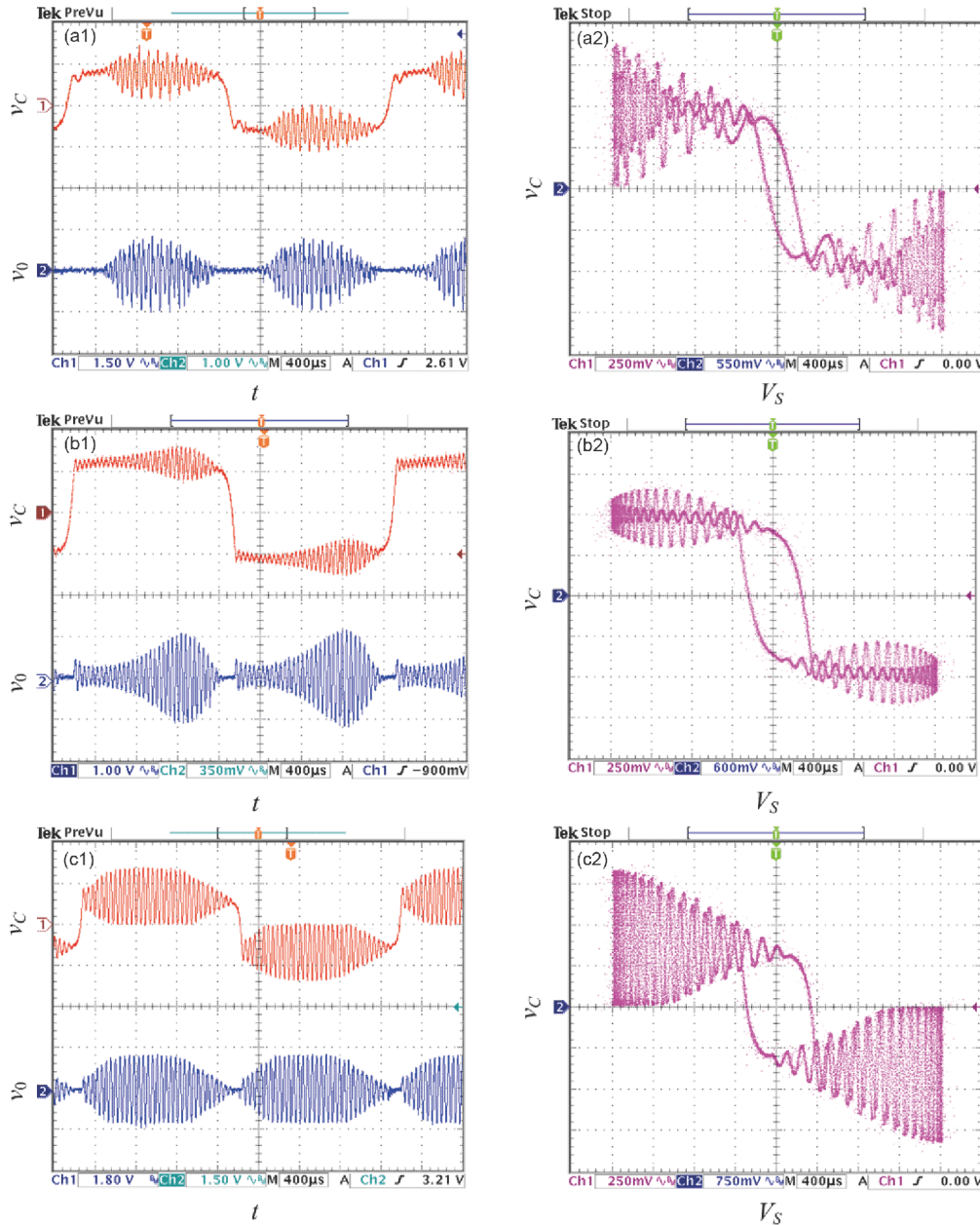


Figure 9 (Color online) Experimentally measured chaotic and bursting activities, where (a1)–(c1) are the time series of v_C and v_0 , (a2)–(c2) are the transformed phase portraits in V_S - v_C plane. (a) $R_1 = -0.82 \text{ k}\Omega$ and $L = 14.47 \text{ mH}$; (b) $R_1 = -0.57 \text{ k}\Omega$ and $L = 3.23 \text{ mH}$; (c) $R_1 = -0.46 \text{ k}\Omega$ and $L = 3.23 \text{ mH}$.

experimental circuit is fabricated. An AD711KN operational amplifier with voltage supplies of ± 15 V linked with three precision potentiometers is used to implement the negative resistor R_1 . Four 1N4148 diodes cascaded with a manually winding inductor and a ceramic capacitor are utilized to construct the second-order memristive diode bridge emulator. The AC voltage source is obtained from a signal generator. The experimental results are measured by a digital oscilloscope.

When the circuit parameters are measured as $V_m=1$ V, $f=0.32$ kHz, $C=9.95$ nF, $C_0=3.21$ nF, $L=14.47$ mH, $L_0=5.3$ mH, $R=1$ k Ω , and $R_1=-0.82$ k Ω , the time series of v_C and v_0 as well as transformed phase portrait in V_S - v_C plane are captured in Figure 9(a). Meanwhile, when two adjustable circuit parameters are determined as $R_1=-0.57$ k Ω and $L=3.23$ mH, the corresponding time series and transformed phase portrait are obtained in Figure 9(b). Furthermore, when the circuit parameters are tuned as $R_1=-0.46$ k Ω and $L=3.23$ mH, the results of the time series and transformed phase portrait are given in Figure 9(c). Compared Figure 9 with Figures 7 and 8, the results in Figure 9(a) agree with those in Figure 7, whereas the results in Figure 9(b) and (c) coincide with those in Figure 8(a) and (b), respectively.

Neglecting the circuit parameter migrations caused by the limited accuracy of element values, parasitic parameters of the experimental circuits, and the measuring errors, the experimental results in Figure 9 can confirm that this non-autonomous memristive FHN circuit indeed exhibits various kinds of bursting activities.

6 Conclusion

In this paper, a memristive non-autonomous FHN circuit was constructed using a second-order memristive diode bridge emulator. The external forcing term was implemented using an AC voltage source to achieve physical realizability. When the exciting term was specified as a single slowly varying variable, periodic and chaotic bursting activities were observed in the memristive FHN circuit. As a non-autonomous system, these bursting activities were innovatively classified based on the time series of the AC equilibrium points and state variables. With this consideration, the bursting activities via Hopf/subHopf, Hopf/Hopf, and Hopf/fold bifurcations were identified in the specified parameter regions. Besides, the experimental measurements were performed to verify the numerically simulated results. The analytical understandings and hardware experimental investigations of the single memristive FHN circuit may contribute to design, investigation and application of artificial neural networks [53,54]. Particularly, from the viewpoint of the AC equilibrium points, qualitative analyses for dynamics in non-autonomous systems can be readily performed. It provides a

novel perspective for related investigations of non-autonomous systems.

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